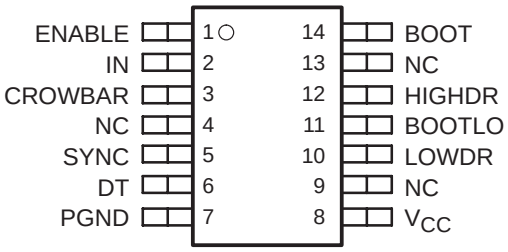
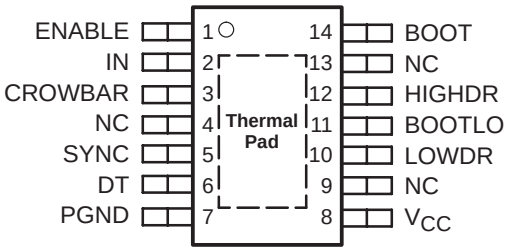
- **Floating Bootstrap or Ground-Reference High-Side Driver**
- **Adaptive Dead-Time Control**
- **50-ns Max Rise/Fall Times and 100-ns Max Propagation Delay − 3.3-nF Load**
- **Ideal for High-Current Single or Multiphase Power Supplies**
- **2.4-A Typical Peak Output Current**
- **4.5-V to 15-V Supply Voltage Range**
- **Internal Schottky Bootstrap Diode**
- **SYNC Control for Synchronous or Nonsynchronous Operation**
- **CROWBAR for OVP, Protects Against Faulted High-Side Power FETs**
- **Low Supply Current....3-mA Typical**
- **−40°C to 125°C Operating Virtual Junction Temperature Range**
- **Available in SOIC and TSSOP PowerPAD Packages**

D PACKAGE
(TOP VIEW)



PWP PACKAGE
(TOP VIEW)



NC − No internal connection

description

The TPS2830 and TPS2831 are MOSFET drivers for synchronous-buck power stages. These devices are ideal for designing a high-performance power supply using switching controllers that do not have MOSFET drivers. The drivers are designed to deliver 2.4-A peak currents into large capacitive loads. The high-side driver can be configured as a ground-reference driver or as a floating bootstrap driver. An adaptive dead-time control circuit eliminates shoot-through currents through the main power FETs during switching transitions, providing higher efficiency for the buck regulator. The TPS2830/31 drivers have additional control functions: ENABLE, SYNC, and CROWBAR. Both drivers are off when ENABLE is low. The driver is configured as a nonsynchronous-buck driver, disabling the low side driver when SYNC is low. The CROWBAR function turns on the low-side power FET, overriding the IN signal, for over-voltage protection against faulted high-side power FETs.

The TPS2830 has a noninverting input. The TPS2831 has an inverting input. The TPS2830/31 drivers are available in 14-terminal SOIC and thermally-enhanced TSSOP PowerPAD™ packages, and operate over a virtual junction temperature range of −40°C to 125°C.

Related Synchronous MOSFET Drivers

DEVICE NAME	ADDITIONAL FEATURES	INPUTS	
TPS2832	W/O ENABLE, SYNC, and CROWBAR	CMOS	Noninverted
TPS2833			Inverted
TPS2834	ENABLE, SYNC, and CROWBAR	TTL	Noninverted
TPS2835			Inverted
TPS2836	W/O ENABLE, SYNC, and CROWBAR	TTL	Noninverted
TPS2837			Inverted

Please be aware that an important notice concerning availability, standard warranty, and use in critical applications of Texas Instruments semiconductor products and disclaimers thereto appears at the end of this data sheet.

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PRODUCTION DATA information is current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.

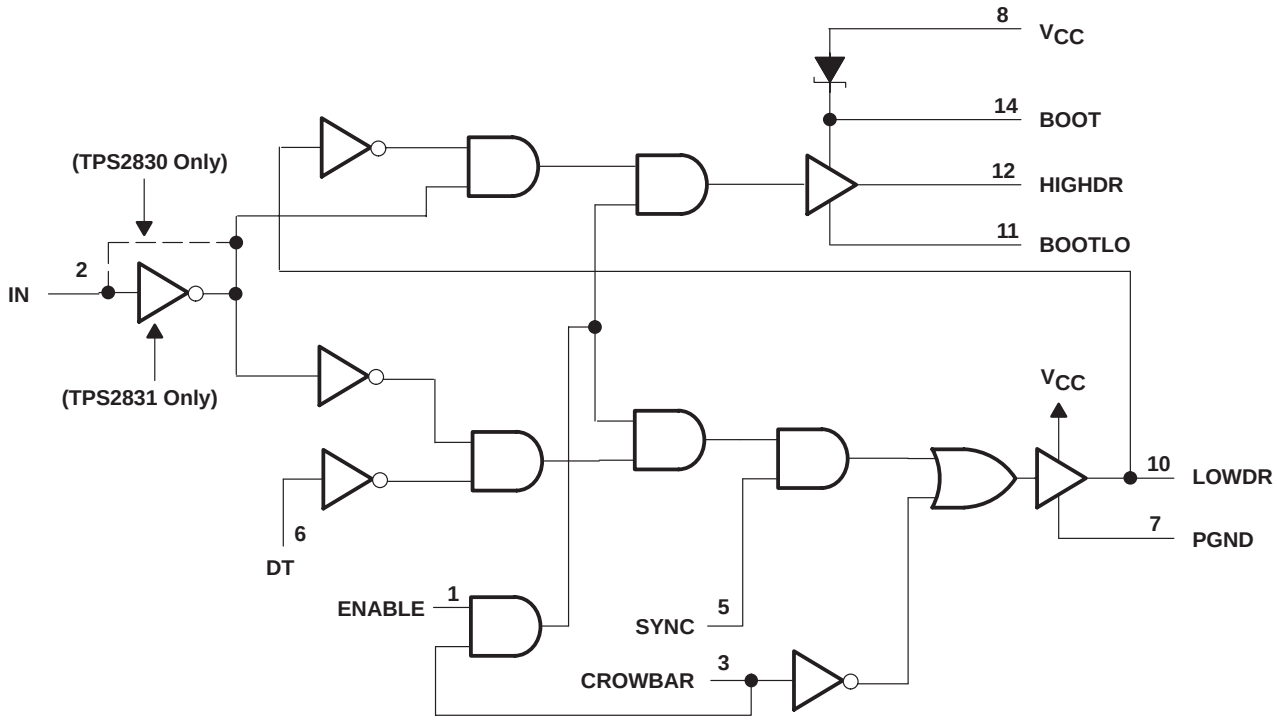
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AVAILABLE OPTIONS

T_J	PACKAGED DEVICES	
	SOIC (D)	TSSOP (PWP)
−40°C to 125°C	TPS2830D TPS2831D	TPS2830PWP TPS2831PWP

The D and PWP packages are available taped and reeled. Add R suffix to device type (e.g., TPS2830DR)

functional block diagram



Terminal Functions

TERMINAL NAME	TERMINAL NO.	I/O	DESCRIPTION
BOOT	14	I	Bootstrap terminal. A ceramic capacitor is connected between BOOT and BOOTLO terminals to develop the floating bootstrap voltage for the high-side MOSFET. The capacitor value is typically between 0.1 μF and 1 μF. A 1-MΩ resistor should be connected across the bootstrap capacitor to provide a discharge path when the driver has been powered down.
BOOTLO	11	O	This terminal connects to the junction of the high-side and low-side MOSFETs.
CROWBAR	3	I	CROWBAR can to be driven by an external OVP circuit to protect against a short across the high-side MOSFET. If CROWBAR is driven low, the low-side driver will be turned on and the high-side driver will be turned off, independent of the status of all other control terminals.
DT	6	I	Dead-time control terminal. Connect DT to the junction of the high-side and low-side MOSFETs.
ENABLE	1	I	If ENABLE is low, both drivers are off.
HIGHDR	12	O	Output drive for the high-side power MOSFET
IN	2	I	Input signal to the MOSFET drivers (noninverting input for the TPS2830; inverting input for the TPS2831).
LOWDR	10	O	Output drive for the low-side power MOSFET
NC	4, 9, 13		No internal connection
PGND	7		Power ground. Connect to the FET power ground
SYNC	5	I	Synchronous Rectifier Enable terminal. If SYNC is low, the low-side driver is always off; If SYNC is high, the low-side driver provides gate drive to the low-side MOSFET.
V_{CC}	8	I	Input supply. Recommended that a 1-μF capacitor be connected from V_{CC} to PGND.

detailed description

low-side driver

The low-side driver is designed to drive low Rds(on) N-channel MOSFETs. The current rating of the driver is 2 A, source and sink.

high-side driver

The high-side driver is designed to drive low Rds(on) N-channel MOSFETs. The current rating of the driver is 2 A, source and sink. The high-side driver can be configured as a GND-reference driver or as a floating bootstrap driver. The internal bootstrap diode is a Schottky, for improved drive efficiency. The maximum voltage that can be applied from BOOT to ground is 30 V.

dead-time (DT) control†

Dead-time control prevents shoot through current from flowing through the main power FETs during switching transitions by controlling the turn-on times of the MOSFET drivers. The high-side driver is not allowed to turn on until the gate drive voltage to the low-side FET is low, and the low-side driver is not allowed to turn on until the voltage at the junction of the power FETs (Vdrain) is low; the DT terminal connects to the junction of the power FETs.

ENABLE†

The ENABLE terminal enables the drivers. When enable is low, the output drivers are low.

IN†

The IN terminal is the input control signal for the drivers. The TPS2830 has a noninverting input; the TPS2831 has an inverting input.

†High-level input voltages on ENABLE, SYNC, CROWBAR, IN, and DT must be greater than or equal to $0.7V_{CC}$.

detailed description (continued)

SYNC†

The SYNC terminal controls whether the drivers operate in synchronous or nonsynchronous mode. In synchronous mode, the low-side FET is operated as a synchronous rectifier. In nonsynchronous mode, the low-side FET is always off.

CROWBAR†

The CROWBAR terminal overrides the normal operation of the driver. When the CROWBAR terminal is low, the low-side FET turns on to act as a clamp, protecting the output voltage of the dc/dc converter against over voltages due to a short across the high-side FET. V_{IN} should be fused to protect the low-side FET.

absolute maximum ratings over operating free-air temperature (unless otherwise noted)†

Supply voltage range, V_{CC} (see Note 1) . . . . . −0.3 V to 16 V
Input voltage range: BOOT to PGND (high-side driver ON) . . . . . −0.3 V to 30 V
BOOTLO to PGND . . . . . −0.3 V to 16 V
BOOT to BOOTLO . . . . . −0.3 V to 16 V
ENABLE, SYNC, and CROWBAR (see Note 2) . . . . . −0.3 V to 16 V
IN (see Note 2) . . . . . −0.3 V to 16 V
DT (see Note 2) . . . . . −0.3 V to 30 V
Continuous total power dissipation . . . . . See Dissipation Rating Table
Operating virtual junction temperature range, T_J . . . . . −40°C to 125°C
Storage temperature range, T_{stg} . . . . . −65°C to 150°C
Lead temperature soldering 1,6 mm (1/16 inch) from case for 10 seconds . . . . . 260°C

‡ Stresses beyond those listed under “absolute maximum ratings” may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under “recommended operating conditions” is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

NOTES: 1. Unless otherwise specified, all voltages are with respect to PGND.
2. High-level input voltages on the ENABLE, SYNC, CROWBAR, IN, and DT terminals must be greater than or equal to $0.7V_{CC}$.

DISSIPATION RATING TABLE

PACKAGE	$T_A \le 25°C$	DERATING FACTOR	$T_A = 70°C$	$T_A = 85°C$
PWP with solder§	2668	26.68 mW/°C	1467	1067
PWP without solder§	1024	10.24 mW/°C	563	409
D	749	7.49 mW/°C	412	300

JUNCTION-CASE THERMAL RESISTANCE TABLE

PWP	Junction-case thermal resistance	2.07 °C/W

§ Test Board Conditions:
1. Thickness: 0.062″
2. 3″ × 3″ (for packages <27 mm long)
3. 4″ × 4″ (for packages >27 mm long)
4. 2 oz copper traces located on the top of the board (0.071 mm thick)
5. Copper areas located on the top and bottom of the PCB for soldering
6. Power and ground planes, 1 oz copper (0.036 mm thick)
7. Thermal vias, 0.33 mm diameter, 1.5 mm pitch
8. Thermal isolation of power plane

For more information, refer to TI technical brief, literature number SLMA002.

†High-level input voltages on ENABLE, SYNC, CROWBAR, IN, and DT must be greater than or equal to $0.7V_{CC}$.

recommended operating conditions

		MIN	NOM	MAX	UNIT
Supply voltage, V_{CC}		4.5		15	V
Input voltage	BOOT to PGND	4.5		28	V

electrical characteristics over recommended operating virtual junction temperature range, V_{CC} = 6.5 V, ENABLE = High, C_L = 3.3 nF (unless otherwise noted)

supply current

PARAMETER		TEST CONDITIONS			MIN	TYP	MAX	UNIT
V_{CC}	Supply voltage range				4.5		15	V
V_{CC}	Quiescent current	V_{ENABLE} = LOW,	V_{CC} =15 V				100	μA
		V_{ENABLE} = HIGH,	V_{CC} =15 V			0.1		mA
		V_{ENABLE} = HIGH, BOOTLO grounded, See Note 3	V_{CC} =12 V, C_{HIGHDR} = 50 pF,	f_{SWX} = 200 kHz, C_{LOWDR} = 50 pF,		3		

NOTE 3: Ensured by design, not production tested.

output drivers

PARAMETER			TEST CONDITIONS		MIN	TYP	MAX	UNIT
Peak output-current	High-side sink (see Note 4)	Duty cycle < 2%, t_{pw} < 100 μs (see Note 3)	$V_{BOOT} - V_{BOOTLO}$ = 4.5 V,	V_{HIGHDR} = 4 V	0.7	1.1		A
			$V_{BOOT} - V_{BOOTLO}$ = 6.5 V,	V_{HIGHDR} = 5 V	1.1	1.5		
			$V_{BOOT} - V_{BOOTLO}$ = 12 V,	V_{HIGHDR} = 10.5 V	2	2.4		
	High-side source (see Note 4)	Duty cycle < 2%, t_{pw} < 100 μs (see Note 3)	$V_{BOOT} - V_{BOOTLO}$ = 4.5 V,	V_{HIGHDR} = 0.5V	1.2	1.4		A
			$V_{BOOT} - V_{BOOTLO}$ = 6.5 V,	V_{HIGHDR} = 1.5 V	1.3	1.6		
			$V_{BOOT} - V_{BOOTLO}$ = 12 V,	V_{HIGHDR} = 1.5 V	2.3	2.7		
	Low-side sink (see Note 4)	Duty cycle < 2%, t_{pw} < 100 μs (see Note 3)	V_{CC} = 4.5 V,	V_{LOWDR} = 4 V	1.3	1.8		A
			V_{CC} = 6.5 V,	V_{LOWDR} = 5 V	2	2.5		
			V_{CC} = 12 V,	V_{LOWDR} = 10.5 V	3	3.5		
	Low-side source (see Note 4)	Duty cycle < 2%, t_{pw} < 100 μs (see Note 3)	V_{CC} = 4.5 V,	V_{LOWDR} = 0.5V	1.4	1.7		A
			V_{CC} = 6.5 V,	V_{LOWDR} = 1.5 V	2	2.4		
			V_{CC} = 12 V,	V_{LOWDR} = 1.5 V	2.5	3		
Output resistance	High-side sink (see Note 4)		$V_{BOOT} - V_{BOOTLO}$ = 4.5 V,	V_{HIGHDR} = 0.5 V			5	Ω
			$V_{BOOT} - V_{BOOTLO}$ = 6.5 V,	V_{HIGHDR} = 0.5 V			5	
			$V_{BOOT} - V_{BOOTLO}$ = 12 V,	V_{HIGHDR} = 0.5 V			5	
	High-side source (see Note 4)		$V_{BOOT} - V_{BOOTLO}$ = 4.5 V,	V_{HIGHDR} = 4 V			75	Ω
			$V_{BOOT} - V_{BOOTLO}$ = 6.5 V,	V_{HIGHDR} = 6 V			75	
			$V_{BOOT} - V_{BOOTLO}$ = 12 V,	V_{HIGHDR} =11.5 V			75	
	Low-side sink (see Note 4)		V_{DRV} = 4.5 V,	V_{LOWDR} = 0.5 V			9	Ω
			V_{DRV} = 6.5 V	V_{LOWDR} = 0.5 V			7.5	
			V_{DRV} = 12 V,	V_{LOWDR} = 0.5 V			6	
	Low-side source (see Note 4)		V_{DRV} = 4.5 V,	V_{LOWDR} = 4 V			75	Ω
			V_{DRV} = 6.5 V,	V_{LOWDR} = 6 V			75	
			V_{DRV} = 12 V,	V_{LOWDR} = 11.5 V			75	

NOTES: 3. Ensured by design, not production tested.
4. The pullup/pulldown circuits of the drivers are bipolar and MOSFET transistors in parallel. The peak output current rating is the combined current from the bipolar and MOSFET transistors. The output resistance is the Rds(on) of the MOSFET transistor when the voltage on the driver output is less than the saturation voltage of the bipolar transistor.

electrical characteristics over recommended operating virtual junction temperature range, V_{CC} = 6.5 V, ENABLE = High, C_L = 3.3 nF (unless otherwise noted) (continued)

dead-time control

PARAMETER			TEST CONDITIONS	MIN	TYP	MAX	UNIT
V_{IH}	High-level input voltage	LOWDR	Over the V_{CC} range (see Note 3)	$0.7V_{CC}$			V
V_{IL}	Low-level input voltage					1	V
V_{IH}	High-level input voltage	DT	Over the V_{CC} range	$0.7V_{CC}$			V
V_{IL}	Low-level input voltage					1	V

NOTE 3: Ensured by design, not production tested.

digital control terminals (IN, CROWBAR, ENABLE, SYNC)

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
V_{IH}	High-level input voltage	Over the V_{CC} range	$0.7V_{CC}$			V
V_{IL}	Low-level input voltage				1	V

switching characteristics over recommended operating virtual junction temperature range, ENABLE = High, C_L = 3.3 nF (unless otherwise noted)

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
Rise time	HIGHDR output (see Note 3)	V_{BOOT} = 4.5 V, V_{BOOTLO} = 0 V			60	ns
		V_{BOOT} = 6.5 V, V_{BOOTLO} = 0 V			50	
		V_{BOOT} = 12 V, V_{BOOTLO} = 0 V			50	
	LOWDR output (see Note 3)	V_{CC} = 4.5 V			40	ns
		V_{CC} = 6.5 V			30	
		V_{CC} = 12 V			30	
Fall time	HIGHDR output (see Note 3)	V_{BOOT} = 4.5 V, V_{BOOTLO} = 0 V			60	ns
		V_{BOOT} = 6.5 V, V_{BOOTLO} = 0 V			50	
		V_{BOOT} = 12 V, V_{BOOTLO} = 0 V			50	
	LOWDR output (see Note 3)	V_{CC} = 4.5 V			40	ns
		V_{CC} = 6.5 V			30	
		V_{CC} = 12 V			30	
Propagation delay time	HIGHDR going low (excluding dead time) (see Note 3)	V_{BOOT} = 4.5 V, V_{BOOTLO} = 0 V			130	ns
		V_{BOOT} = 6.5 V, V_{BOOTLO} = 0 V			100	
		V_{BOOT} = 12 V, V_{BOOTLO} = 0 V			75	
	LOWDR going high (excluding dead time) (see Note 3)	V_{BOOT} = 4.5 V, V_{BOOTLO} = 0 V			80	ns
		V_{BOOT} = 6.5 V, V_{BOOTLO} = 0 V			70	
		V_{BOOT} = 12 V, V_{BOOTLO} = 0 V			60	
Propagation delay time	LOWDR going low (excluding dead time) (see Note 3)	V_{CC} = 4.5 V			80	ns
		V_{CC} = 6.5 V			70	
		V_{CC} = 12 V			60	
Driver nonoverlap time	DT to LOWDR and LOWDR to HIGHDR (see Note 3)	V_{CC} = 4.5 V	40		170	ns
		V_{CC} = 6.5 V	25		135	
		V_{CC} = 12 V	15		85	

NOTE 3: Ensured by design, not production tested.

TYPICAL CHARACTERISTICS

RISE TIME
vs
SUPPLY VOLTAGE

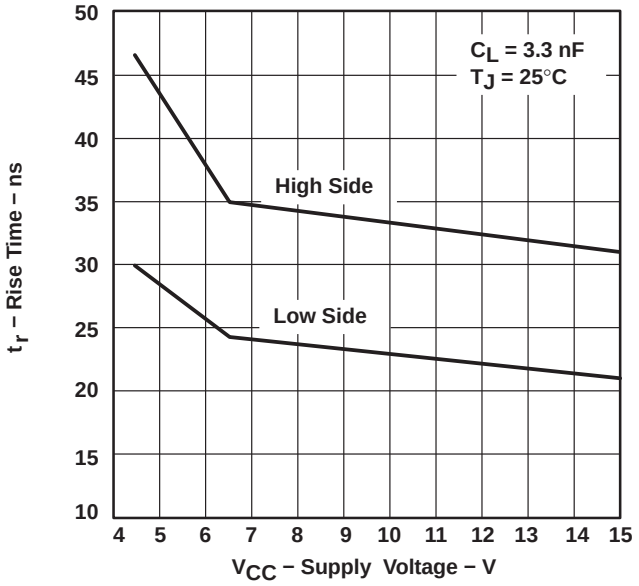


Figure 1

FALL TIME
vs
SUPPLY VOLTAGE

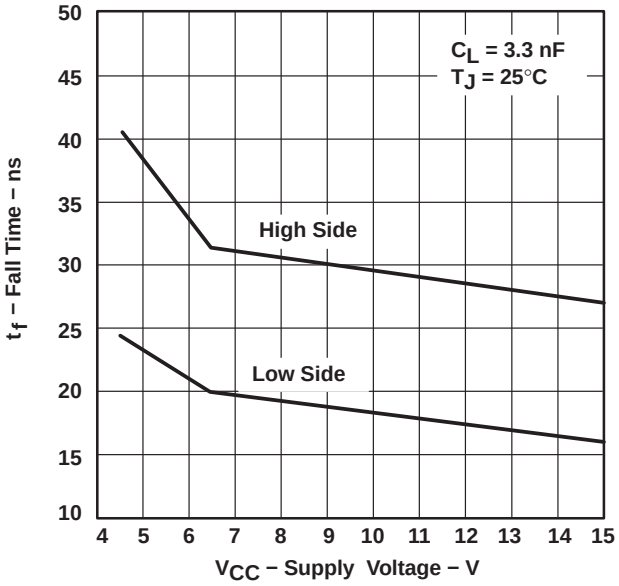


Figure 2

RISE TIME
vs
JUNCTION TEMPERATURE

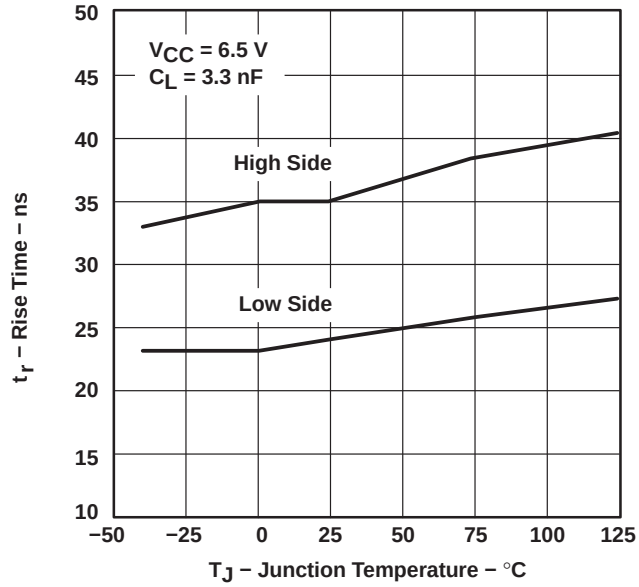


Figure 3

FALL TIME
vs
JUNCTION TEMPERATURE

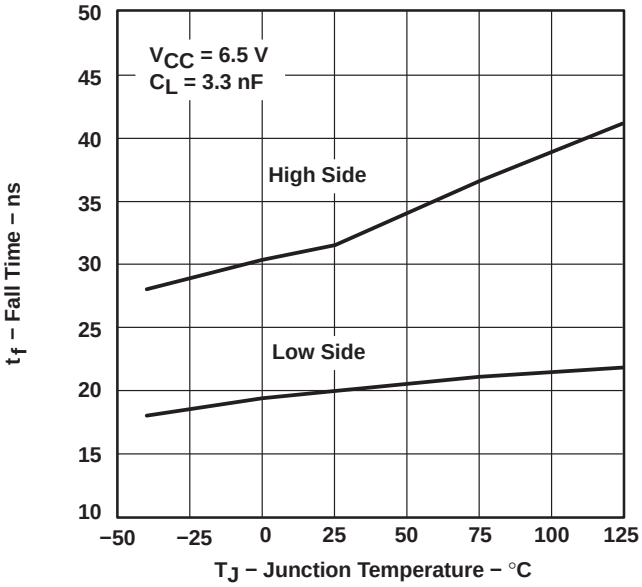


Figure 4

TYPICAL CHARACTERISTICS

LOW-TO-HIGH PROPAGATION DELAY TIME
vs
SUPPLY VOLTAGE, LOW TO HIGH LEVEL

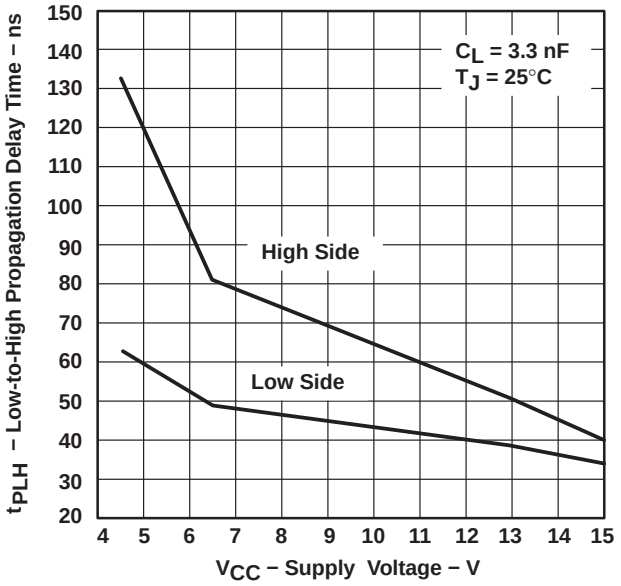


Figure 5

HIGH-TO-LOW PROPAGATION DELAY TIME
vs
SUPPLY VOLTAGE, HIGH TO LOW LEVEL

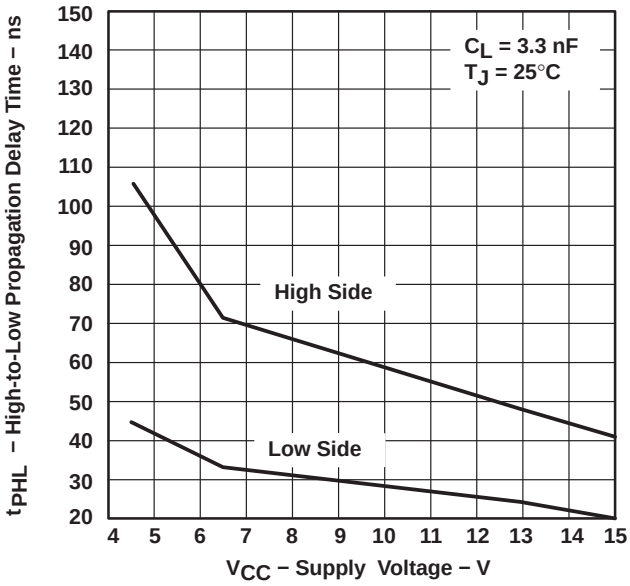


Figure 6

LOW-TO-HIGH PROPAGATION DELAY TIME
vs
JUNCTION TEMPERATURE

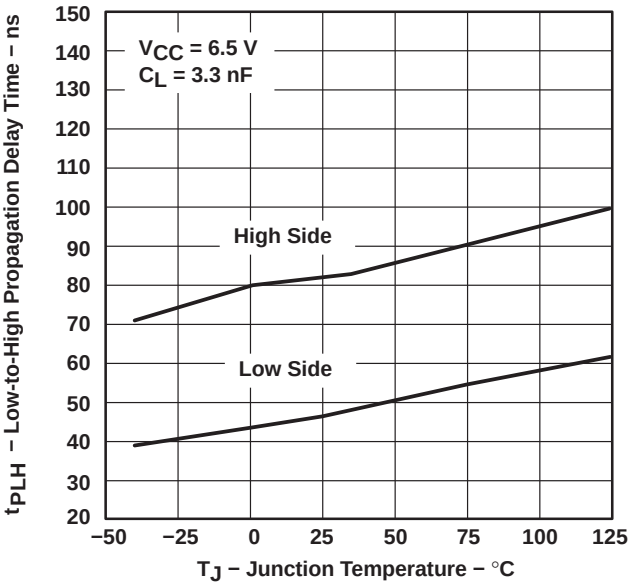


Figure 7

HIGH-TO-LOW PROPAGATION DELAY TIME
vs
JUNCTION TEMPERATURE

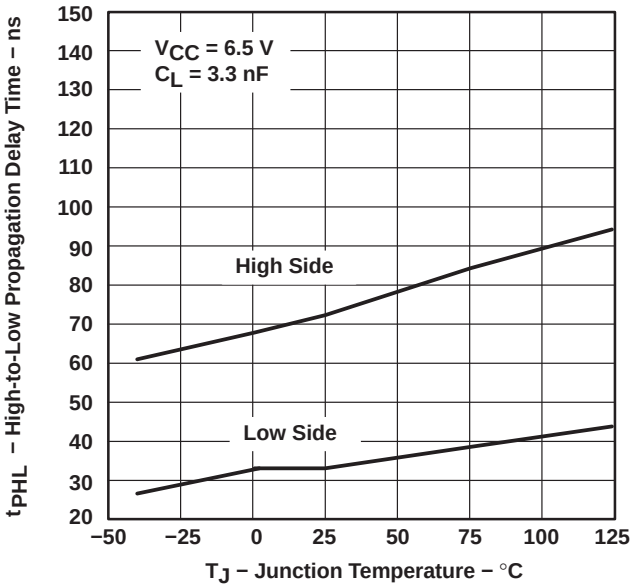


Figure 8

TYPICAL CHARACTERISTICS

RISE TIME
vs
LOAD CAPACITANCE

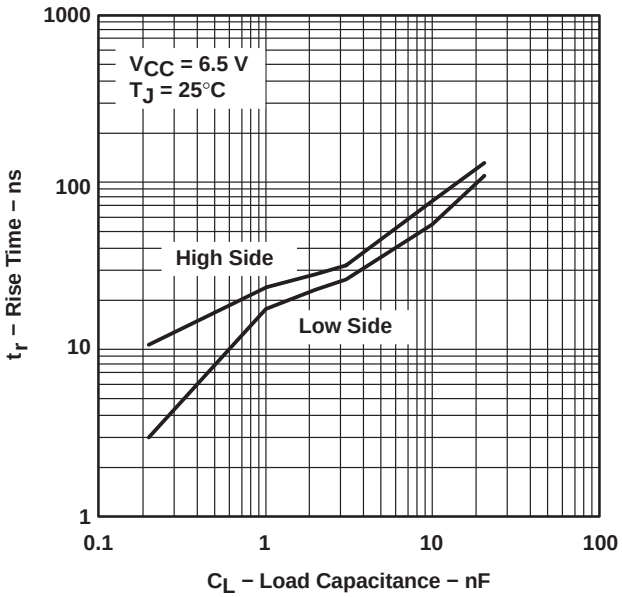


Figure 9

FALL TIME
vs
LOAD CAPACITANCE

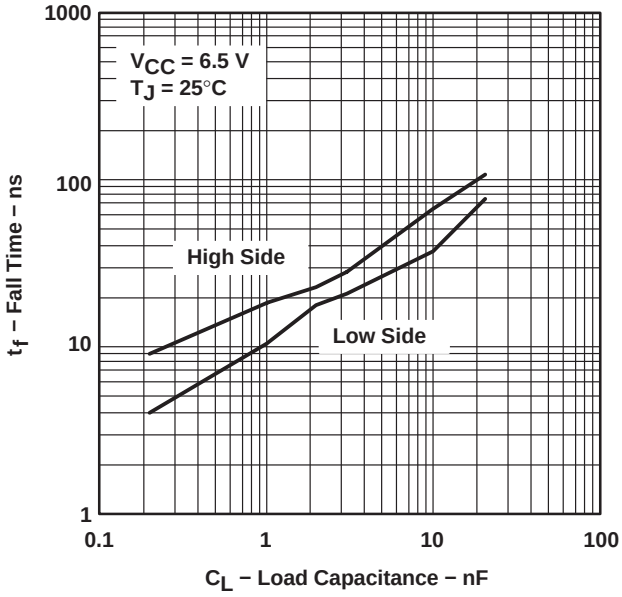


Figure 10

SUPPLY CURRENT
vs
SUPPLY VOLTAGE

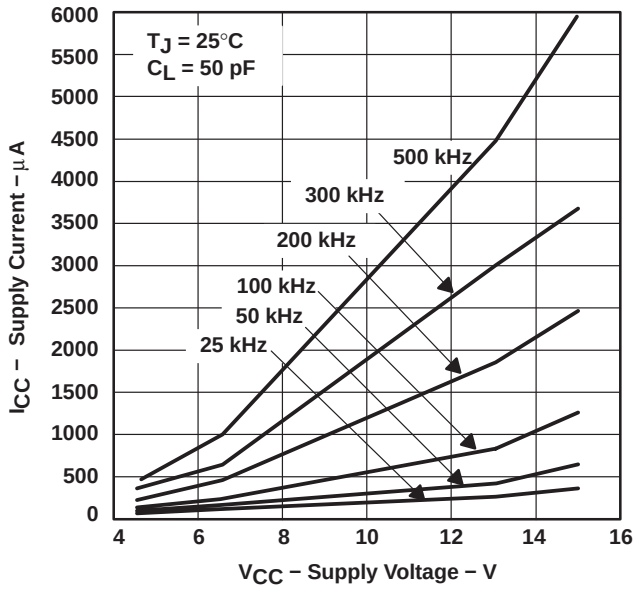


Figure 11

SUPPLY CURRENT
vs
SUPPLY VOLTAGE

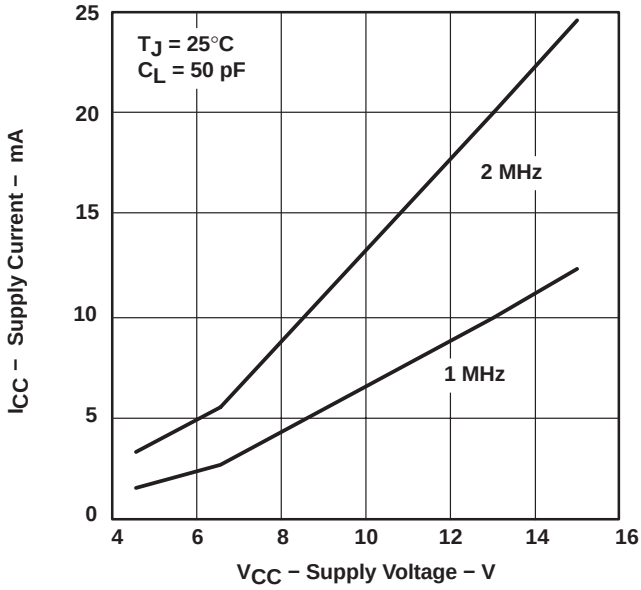


Figure 12

TYPICAL CHARACTERISTICS

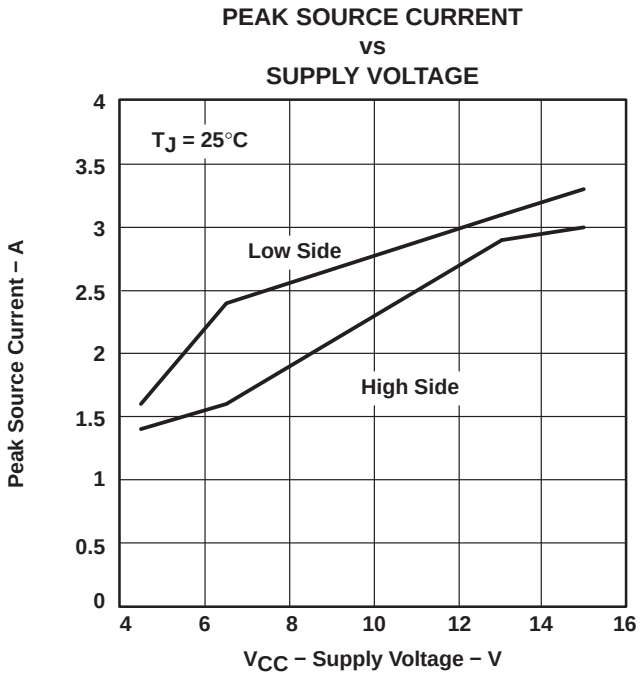


Figure 13

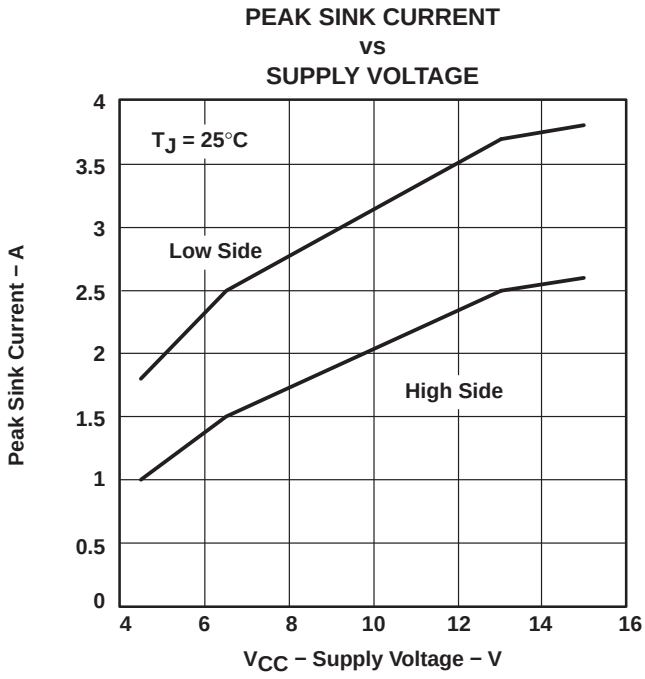


Figure 14

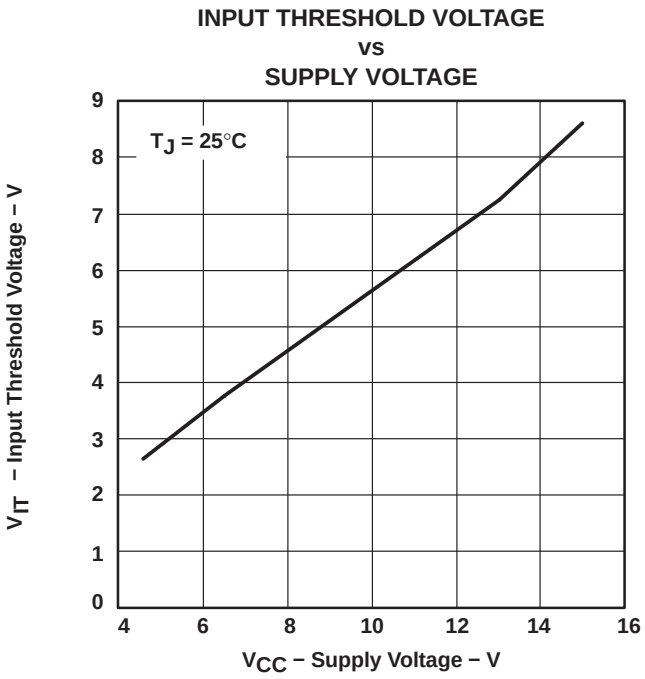


Figure 15

APPLICATION INFORMATION

Figure 16 shows the circuit schematic of a 100-kHz synchronous-buck converter implemented with a TL5001A pulse-width-modulation (PWM) controller and a TPS2831 driver. The converter operates over an input range from 4.5 V to 12 V and has a 3.3-V output. The circuit can supply 3 A continuous load. The converter achieves an efficiency of 94% for V_{IN} = 5 V, I_{load}=1 A, and 93% for V_{in} = 5 V, I_{load} = 3 A.

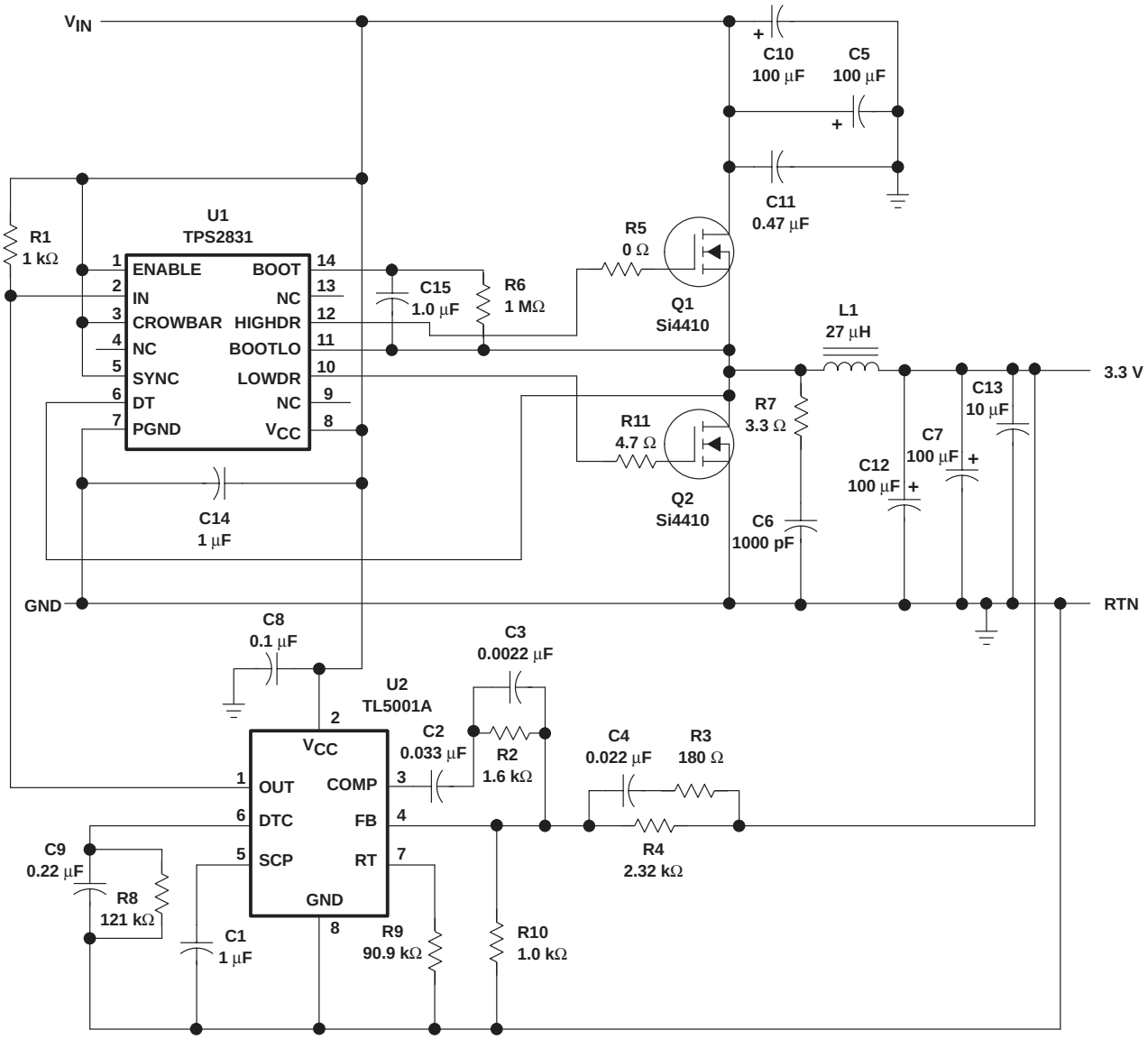


Figure 16. 3.3-V 3-A Synchronous-Buck Converter Circuit

APPLICATION INFORMATION

Great care should be taken when laying out the PC board. The power-processing section is the most critical and will generate large amounts of EMI if not properly configured. The junction of Q1, Q2, and L1 should be very tight. The connection from Q1 drain to the positive sides of C5, C10, and C11 and the connection from Q2 source to the negative sides of C5, C10, and C11 should be as short as possible. The negative terminals of C7 and C12 should also be connected to Q2 source.

Next, the traces from the MOSFET driver to the power switches should be considered. The BOOTLO signal from the junction of Q1 and Q2 carries the large gate drive current pulses and should be as heavy as the gate drive traces. The bypass capacitor (C14) should be tied directly across V_{CC} and PGND.

The next most sensitive node is the FB node on the controller (terminal 4 on the TL5001A) This node is very sensitive to noise pickup and should be isolated from the high-current power stage and be as short as possible. The ground around the controller and low-level circuitry should be tied to the power ground as the output. If these three areas are properly laid out, the rest of the circuit should not have any other EMI problems and the power supply will be relatively free of noise.

PACKAGING INFORMATION

Orderable Device	Status (1)	Package Type	Package Drawing	Pins	Package Qty	Eco Plan (2)	Lead/ Ball Finish	MSL Peak Temp (3)	Samples (Requires Login)
TPS2830D	ACTIVE	SOIC	D	14	50	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	
TPS2830DG4	ACTIVE	SOIC	D	14	50	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	
TPS2830PWP	ACTIVE	HTSSOP	PWP	14	90	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	
TPS2830PWPG4	ACTIVE	HTSSOP	PWP	14	90	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	
TPS2830PWPR	ACTIVE	HTSSOP	PWP	14	2000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	
TPS2830PWPRG4	ACTIVE	HTSSOP	PWP	14	2000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	
TPS2831D	ACTIVE	SOIC	D	14	50	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	
TPS2831DG4	ACTIVE	SOIC	D	14	50	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	
TPS2831DR	ACTIVE	SOIC	D	14	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	
TPS2831DRG4	ACTIVE	SOIC	D	14	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	
TPS2831PWPR	ACTIVE	HTSSOP	PWP	14	2000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	
TPS2831PWPRG4	ACTIVE	HTSSOP	PWP	14	2000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	

(1) The marketing status values are defined as follows:
ACTIVE: Product device recommended for new designs.
LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.
NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.
PREVIEW: Device has been announced but is not in production. Samples may or may not be available.
OBSOLETE: TI has discontinued the production of the device.

(2) Eco Plan - The planned eco-friendly classification: Pb-Free (RoHS), Pb-Free (RoHS Exempt), or Green (RoHS & no Sb/Br) - please check http://www.ti.com/productcontent for the latest availability information and additional product content details.
TBD: The Pb-Free/Green conversion plan has not been defined.

Pb-Free (RoHS): TI's terms "Lead-Free" or "Pb-Free" mean semiconductor products that are compatible with the current RoHS requirements for all 6 substances, including the requirement that lead not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, TI Pb-Free products are suitable for use in specified lead-free processes.
Pb-Free (RoHS Exempt): This component has a RoHS exemption for either 1) lead-based flip-chip solder bumps used between the die and package, or 2) lead-based die adhesive used between the die and leadframe. The component is otherwise considered Pb-Free (RoHS compatible) as defined above.
Green (RoHS & no Sb/Br): TI defines "Green" to mean Pb-Free (RoHS compatible), and free of Bromine (Br) and Antimony (Sb) based flame retardants (Br or Sb do not exceed 0.1% by weight in homogeneous material)

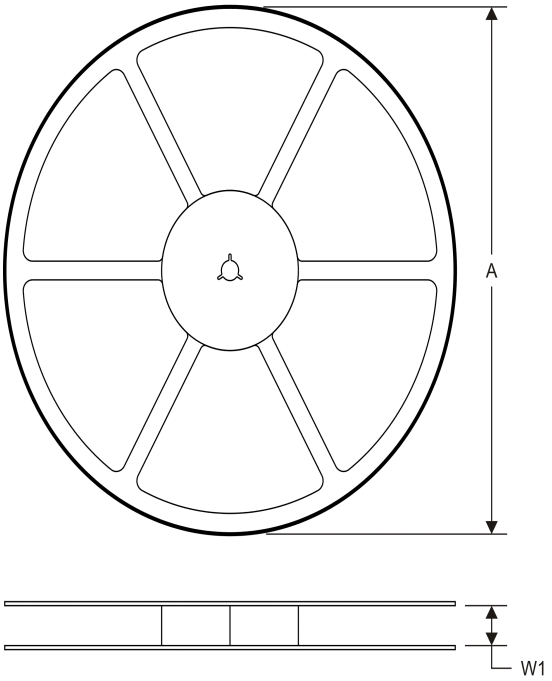
(3) MSL, Peak Temp. -- The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

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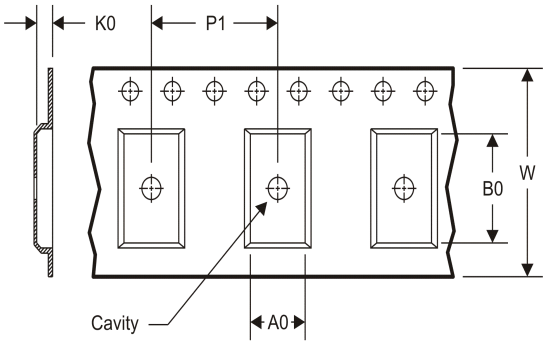
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TAPE AND REEL INFORMATION

REEL DIMENSIONS



TAPE DIMENSIONS



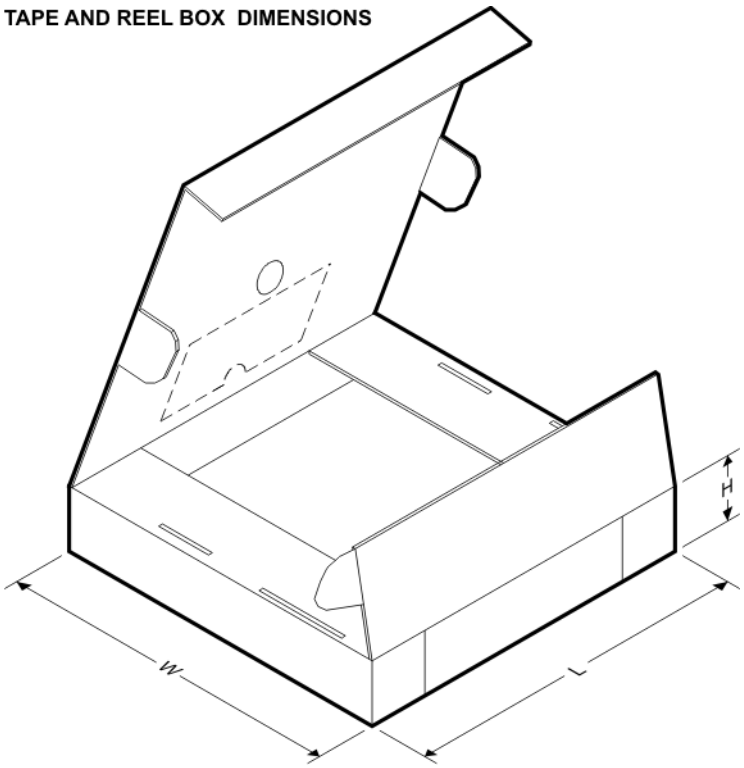
A0	Dimension designed to accommodate the component width
B0	Dimension designed to accommodate the component length
K0	Dimension designed to accommodate the component thickness
W	Overall width of the carrier tape
P1	Pitch between successive cavity centers

TAPE AND REEL INFORMATION

*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
TPS2830PWPR	HTSSOP	PWP	14	2000	330.0	12.4	6.9	5.6	1.6	8.0	12.0	Q1
TPS2831DR	SOIC	D	14	2500	330.0	16.4	6.5	9.0	2.1	8.0	16.0	Q1
TPS2831PWPR	HTSSOP	PWP	14	2000	330.0	12.4	6.9	5.6	1.6	8.0	12.0	Q1

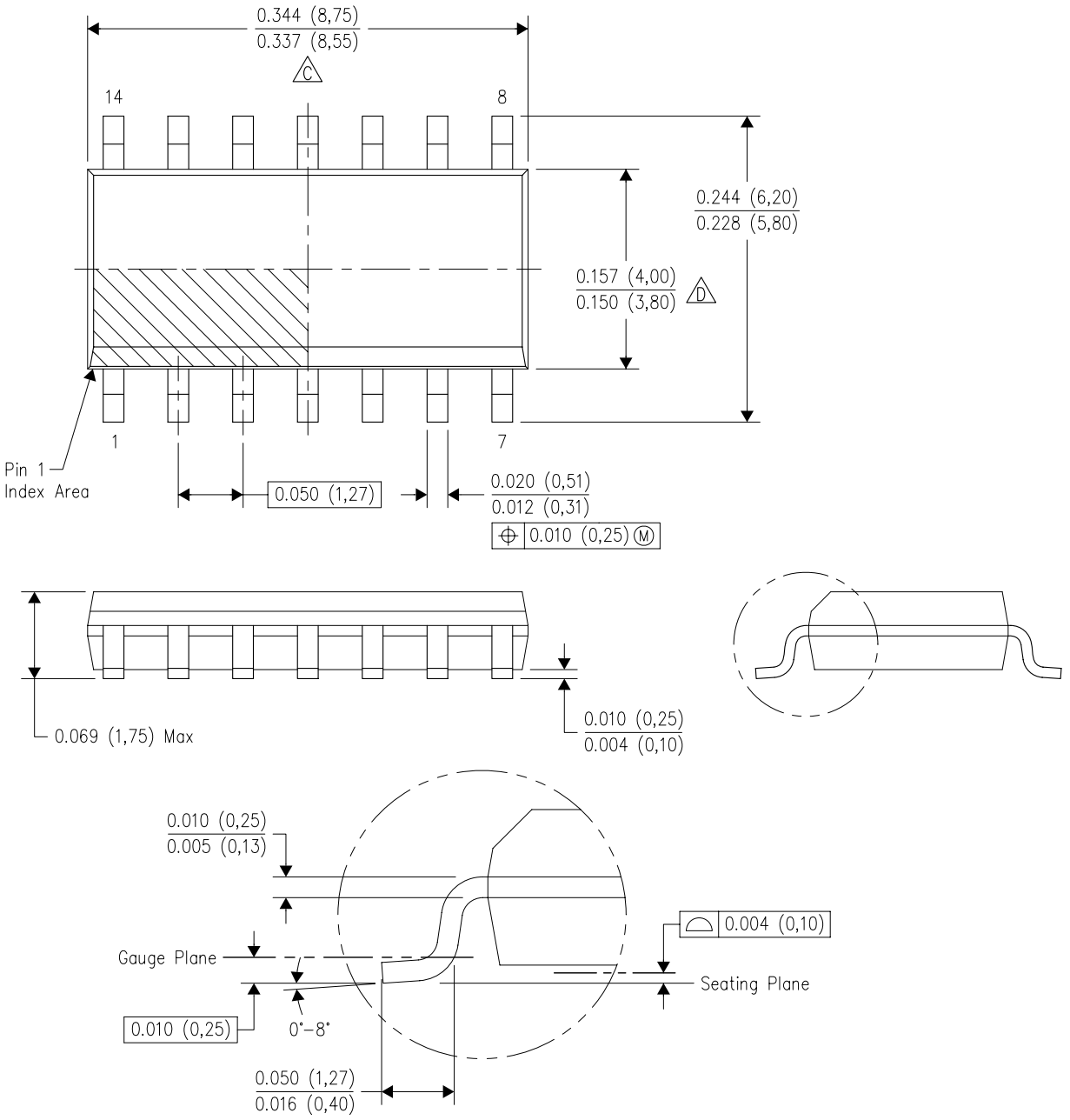
TAPE AND REEL BOX DIMENSIONS



*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
TPS2830PWPR	HTSSOP	PWP	14	2000	367.0	367.0	35.0
TPS2831DR	SOIC	D	14	2500	367.0	367.0	38.0
TPS2831PWPR	HTSSOP	PWP	14	2000	367.0	367.0	35.0

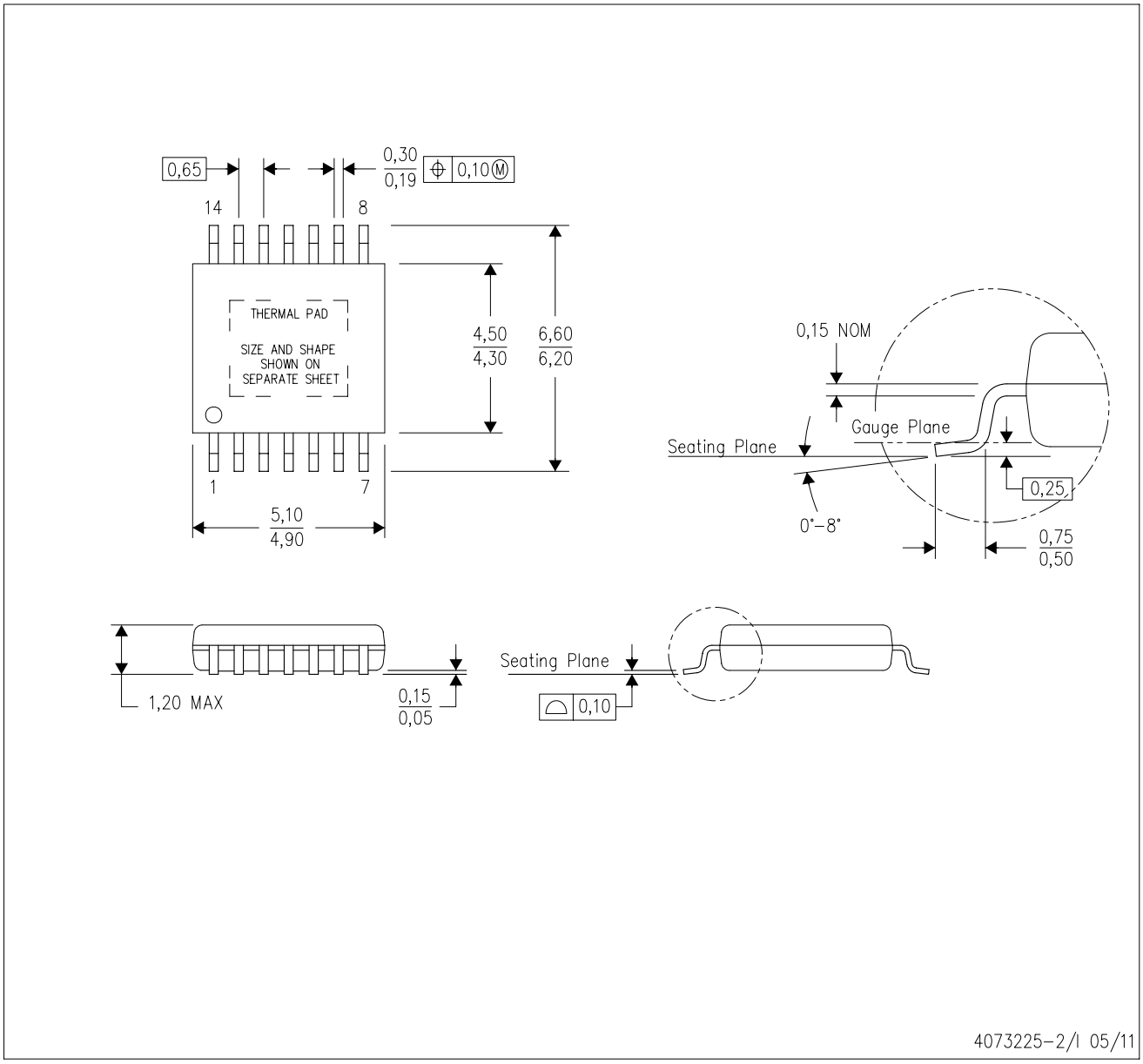
D (R–PDSO–G14) PLASTIC SMALL OUTLINE



4040047–5/M 06/11

NOTES:
A. All linear dimensions are in inches (millimeters).
B. This drawing is subject to change without notice.
C. Body length does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed 0.006 (0,15) each side.
D. Body width does not include interlead flash. Interlead flash shall not exceed 0.017 (0,43) each side.
E. Reference JEDEC MS–012 variation AB.

PWP (R–PDSO–G14) PowerPAD™ PLASTIC SMALL OUTLINE



NOTES:
A. All linear dimensions are in millimeters.
B. This drawing is subject to change without notice.
C. Body dimensions do not include mold flash or protrusions. Mold flash and protrusion shall not exceed 0.15 per side.
D. This package is designed to be soldered to a thermal pad on the board. Refer to Technical Brief, PowerPad Thermally Enhanced Package, Texas Instruments Literature No. SLMA002 for information regarding recommended board layout. This document is available at www.ti.com <http://www.ti.com>.
E. See the additional figure in the Product Data Sheet for details regarding the exposed thermal pad features and dimensions.
E. Falls within JEDEC MO–153

PowerPAD is a trademark of Texas Instruments.

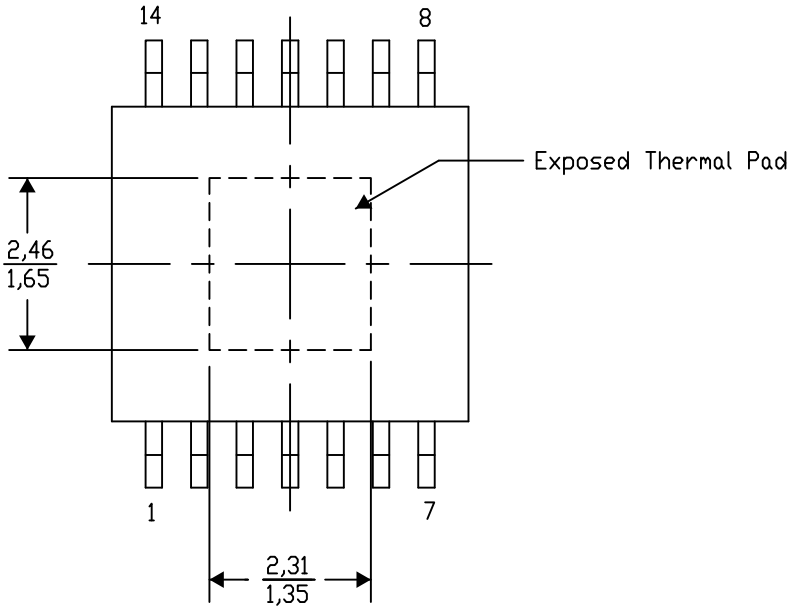
PWP (R–PDSO–G14) PowerPAD™ SMALL PLASTIC OUTLINE

THERMAL INFORMATION

This PowerPAD™ package incorporates an exposed thermal pad that is designed to be attached to a printed circuit board (PCB). The thermal pad must be soldered directly to the PCB. After soldering, the PCB can be used as a heatsink. In addition, through the use of thermal vias, the thermal pad can be attached directly to the appropriate copper plane shown in the electrical schematic for the device, or alternatively, can be attached to a special heatsink structure designed into the PCB. This design optimizes the heat transfer from the integrated circuit (IC).

For additional information on the PowerPAD package and how to take advantage of its heat dissipating abilities, refer to Technical Brief, PowerPAD Thermally Enhanced Package, Texas Instruments Literature No. SLMA002 and Application Brief, PowerPAD Made Easy, Texas Instruments Literature No. SLMA004. Both documents are available at www.ti.com.

The exposed thermal pad dimensions for this package are shown in the following illustration.



Top View

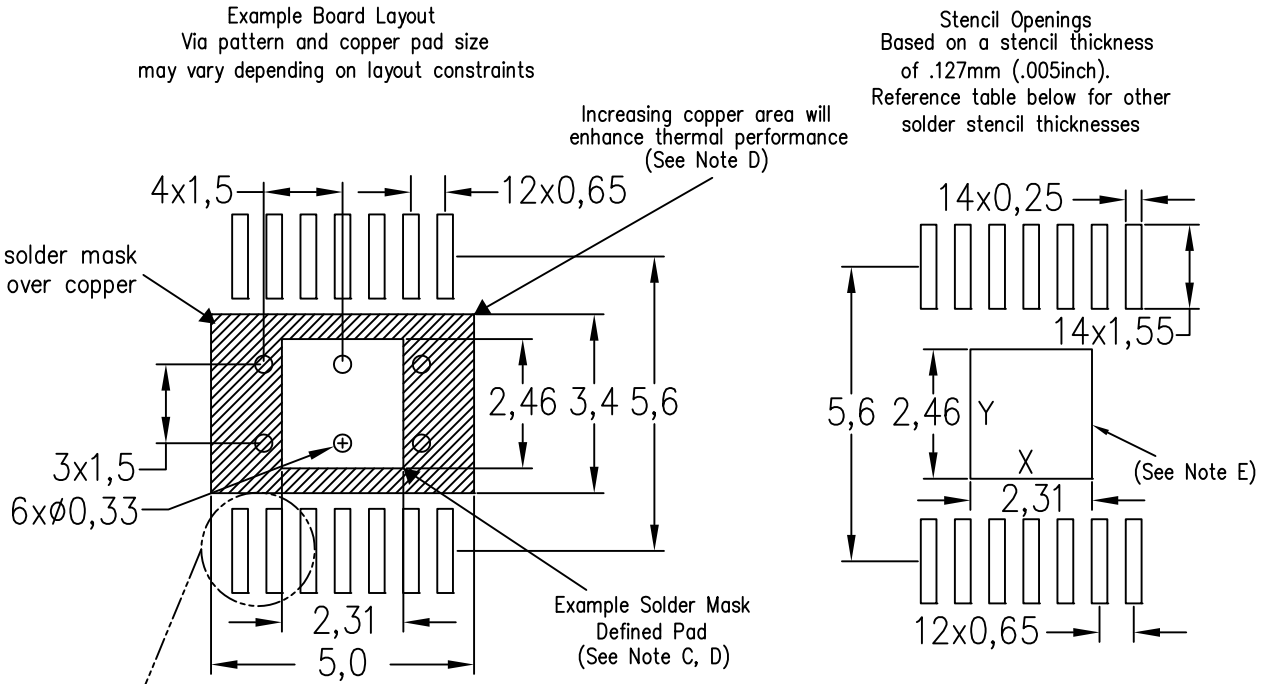
Exposed Thermal Pad Dimensions

4206332–2/AC 07/12

NOTE: A. All linear dimensions are in millimeters

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PWP (R–PDSO–G14) PowerPAD™ PLASTIC SMALL OUTLINE



Center Power Pad Solder Stencil Opening		
Stencil Thickness	X	Y
0.1mm	2.5	2.65
0.127mm	2.31	2.46
0.152mm	2.15	2.3
0.178mm	2.05	2.15

4207609–2/R 07/12

NOTES:
A. All linear dimensions are in millimeters.
B. This drawing is subject to change without notice.
C. Customers should place a note on the circuit board fabrication drawing not to alter the center solder mask defined pad.
D. This package is designed to be soldered to a thermal pad on the board. Refer to Technical Brief, PowerPad Thermally Enhanced Package, Texas Instruments Literature No. SLMA002, SLMA004, and also the Product Data Sheets for specific thermal information, via requirements, and recommended board layout. These documents are available at www.ti.com <http://www.ti.com>. Publication IPC–7351 is recommended for alternate designs.
E. Laser cutting apertures with trapezoidal walls and also rounding corners will offer better paste release. Customers should contact their board assembly site for stencil design recommendations. Example stencil design based on a 50% volumetric metal load solder paste. Refer to IPC–7525 for other stencil recommendations.
F. Customers should contact their board fabrication site for solder mask tolerances between and around signal pads.

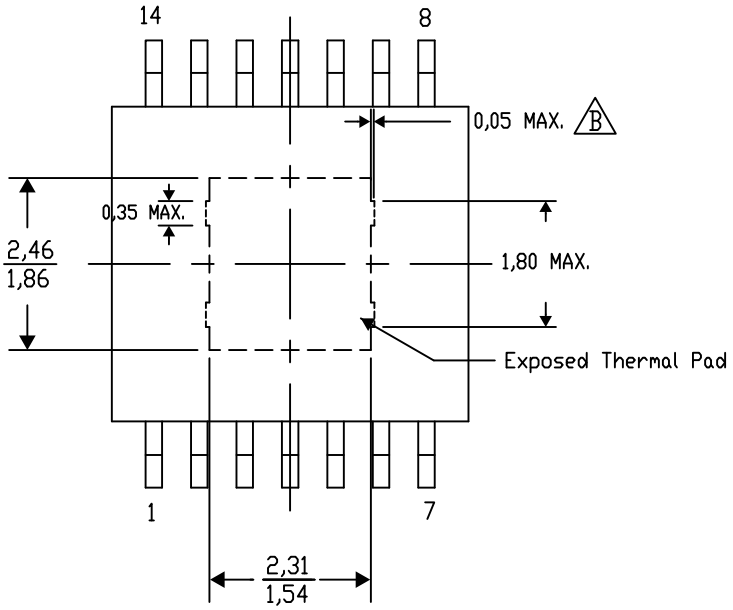
PWP (R–PDSO–G14) PowerPAD™ SMALL PLASTIC OUTLINE

THERMAL INFORMATION

This PowerPAD™ package incorporates an exposed thermal pad that is designed to be attached to a printed circuit board (PCB). The thermal pad must be soldered directly to the PCB. After soldering, the PCB can be used as a heatsink. In addition, through the use of thermal vias, the thermal pad can be attached directly to the appropriate copper plane shown in the electrical schematic for the device, or alternatively, can be attached to a special heatsink structure designed into the PCB. This design optimizes the heat transfer from the integrated circuit (IC).

For additional information on the PowerPAD package and how to take advantage of its heat dissipating abilities, refer to Technical Brief, PowerPAD Thermally Enhanced Package, Texas Instruments Literature No. SLMA002 and Application Brief, PowerPAD Made Easy, Texas Instruments Literature No. SLMA004. Both documents are available at www.ti.com.

The exposed thermal pad dimensions for this package are shown in the following illustration.



Top View

Exposed Thermal Pad Dimensions

4206332–44/AC 07/12

NOTE: A. All linear dimensions are in millimeters

B. Exposed tie strap features may not be present.

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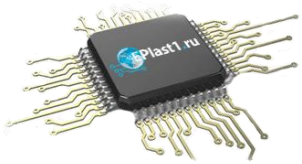
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Телефон: 8 (812) 309 58 32 (многоканальный)
Факс: 8 (812) 320-02-42
Электронная почта: org@eplast1.ru
Адрес: 198099, г. Санкт-Петербург, ул. Калинина, дом 2, корпус 4, литера А.